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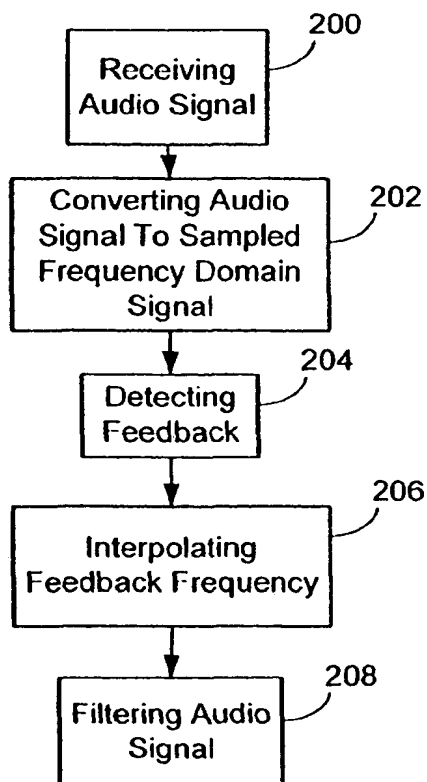
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[Continued on next page]

(54) Title: AUDIO FEEDBACK PROCESSING SYSTEM



(57) Abstract: This invention is an audio system capable of identifying frequencies of feedback signals and filters these feedback signals. Frequency interpolation is utilized on a sampled frequency spectrum signal corresponding to a feedback signal, where the frequency interpolation allows the frequency of the feedback signal to be identified, especially where the frequency lies between samples of the frequency spectrum signal. The frequency identification allows a less intrusive filter, such as a notch filter, to be placed at the determined frequency of the feedback to eliminate the feedback signal. The placement of the notch filter reduces the effect on the audio signal provided by the audio system. The audio system may provide adaptive filtering of multiple feedback signals, using a single filter such as a notch filter. The single notch filter is configured to filter the more than one feedback signal, thereby freeing-up notch filters for filtering other feedback signals, or for reducing memory and processing requirements for the processor of the system.

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## AUDIO FEEDBACK PROCESSING SYSTEM

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### BACKGROUND OF THE INVENTION

#### 10 1. Related Applications.

[01] This application claims priority to U.S. Provisional Patent Application Serial No. 60/363,994, entitled "Employing Narrow Bandwidth Notch Filters In Feedback Elimination," and filed on March 13, 2002, and is incorporated by reference.

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#### 2. Technical Field.

[02] This invention generally relates to feedback in audio systems. More particularly, this invention relates to identifying a frequency of feedback and adaptive filtering of feedback signals in an audio system.

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#### 3. Related Art.

[03] An audio system typically includes an input transducer (microphone), an amplifier, a microprocessor and an audio output (loudspeaker). The input transducer receives sound into the system, the amplifier amplifies the sound, the microprocessor performs signal processing, and the audio output (loudspeaker) provides sound to users of the system. Many audio systems allow for a duplex operation, where sound may be input to the microphone while audio is provided at the speaker. However, when the microphone receives a portion of the audio provided at the speaker as an input, an unstable, closed-loop system is created, resulting in audio feedback.

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[04] Audio feedback is manifested as one or more audio feedback signals at the speaker, where each feedback signal may be modeled as a sinusoidal signal (i.e. the feedback signal(s) exhibit characteristics of a sinusoidal signal). To eliminate a

particular feedback signal, the microprocessor converts the audio signal into a discrete (sampled) frequency spectrum representation, such as a Discrete Fourier Transform (DFT), Spectral Estimation, Filter Banks, or like representation. The conversion of the audio signal to the sampled frequency spectrum allows for a  
5 general identification of the frequency of the feedback signal. The frequency sample having the greatest magnitude in the discrete frequency domain is selected as the frequency of the feedback signal.

[05] A notch filter is placed at the identified frequency of the feedback signal to eliminate that particular feedback signal. However, because of computational and  
10 memory limitations of the microprocessor, the sampling resolution of the sampled frequency spectrum representation is limited. Thus, the selected frequency sample does not provide an accurate estimate of an actual frequency of the feedback signal. Because the selected frequency sample is not an accurate estimate, a notch filter is utilized that has a significantly wider bandwidth and/or a greater cut-depth than what  
15 is actually necessary for filtering the feedback signal. The wider bandwidth and/or greater cut-depth are necessary to ensure that the feedback signal is eliminated from the output signal. However, the use of a wider bandwidth and/or greater cut-depth notch filter can degrade the audio quality of the sound at the speaker.

[06] The computational and memory limitations of the microprocessor limits  
20 the number of notch filters that may be used to eliminate audio feedback signals. Where the number of feedback signals exceeds the number of notch filters available, some of the feedback signals cannot be eliminated by the system. The failure to eliminate at least some of the feedback signals may require a system gain to be reduced, resulting in degraded system performance.

### SUMMARY

[07] This invention provides an audio system that identifies the frequency of a feedback signal using interpolative feedback identification. The interpolative feedback identification may be accomplished using frequency interpolation on a sampled frequency spectrum signal corresponding to a feedback signal. The feedback interpolation allows the frequency of the feedback signal to be identified, especially where the frequency of the feedback lies between samples of the frequency spectrum signal. The interpolation may include using samples of the sampled frequency spectrum signal to generate a unique quadratic (or higher order polynominal), which resembles the original main lobe of the feedback signal represented by the frequency spectrum signal. The polynominal may be constructed from the samples using polynominal interpolation, rational function interpolation, cubic spline interpolation, and the like. The peak of the polynominal and thus a representation/estimation of the actual frequency of the feedback signal may be determined, for example, by setting a derivative of the generated polynominal equation to zero. A narrowly tailored filter, such as a notch filter, may be placed at the determined frequency of the feedback to eliminate or reduce the feedback signal. The filter also reduces the effect on the audio signal quality provided by the audio system.

[08] The audio system may adaptively filter multiple feedback signals using a single filter such as a notch filter. The adaptive filtering may include identifying frequencies of feedback in the audio signal, and determining which frequencies of feedback signals lie within a frequency window comprising adjoining samples of the sampled frequency spectrum. A filter, such as a notch filter is configured to filter out the frequencies identified as within a frequency range covered by the frequency window, thereby freeing-up notch filters for filtering other feedback signals, or for reducing memory and processing requirements for the microprocessor of the audio system. The frequency range covered by the frequency window may comprise any number of adjoining samples, and may be predetermined and/or configurable. Further, the frequency range covered by the frequency window may vary depending

on the frequency band being examined, and/or the resolution of the sampled frequency spectrum.

[09] Other systems, methods, features and advantages of the invention will be, or will become, apparent to one with skill in the art upon examination of the following figures and detailed description. It is intended that all such additional systems, methods, features and advantages be included within this description, be within the scope of the invention, and be protected by the following claims.

#### BRIEF DESCRIPTION OF THE DRAWINGS

[10] The invention can be better understood with reference to the following drawings and description. The components in the figures are not necessarily to scale, emphasis instead being placed upon illustrating the principles of the invention. Moreover, in the figures, like reference numerals designate corresponding parts throughout the different views.

[11] Figure 1 is a block diagram of an audio system having feedback identification and reduction techniques.

[12] Figure 2 is a flow chart illustrating operation of the audio system of Figure 1 in identifying the frequency of a feedback signal.

[13] Figure 3 is a graph illustrating a time-domain feedback signal.

[14] Figure 4 is a graph illustrating the Discrete Time Fourier Transform of the feedback signal of Figure 3.

[15] Figure 5 is a graph illustrating a time-domain window function.

[16] Figure 6 is a graph illustrating a Discrete Time Fourier Transform of the time-domain window function of Figure 5.

[17] Figure 7 is a graph illustrating the time-domain signal resulting from multiplying the feedback signal of Figure 3 with the window function of Figure 5.

[18] Figure 8 is a graph illustrating the Discrete Time Fourier Transform of the windowed feedback signal of Figure 7.

[19] Figure 9 is a graph illustrating the Discrete Fourier Transform of the of the windowed feedback signal of Figure 7.

[20] Figure 10 illustrates an expansion of a portion of the graph of Figure 9, showing frequency bins which may be utilized in interpolating a frequency of a feedback signal.

[21] Figure 11 is a graph comparing characteristics of prior art notch filters with a notch filter configured using interpolative feedback identification.

[22] Figure 12 is another graph comparing characteristics of a prior art notch filter, with a notch filter configured using interpolative feedback identification.

[23] Figure 13 is a flow chart illustrating operation of the audio system of Figure 1 for performing adaptive filtering.

10 [24] Figure 14 is a graph illustrating a frequency window covering a specified frequency range for a time-domain signal, which may be utilized in performing adaptive filtering.

[25] Figure 15 is a graph illustrating a frequency window covering a specified frequency range for a frequency-domain signal, which may be utilized in performing adaptive filtering.

[26] Figure 16 is a graph illustrating characteristics for two notch filters for filtering corresponding feedback signals.

[27] Figure 17 is a graph illustrating characteristics of a notch filter configured for adaptively filtering two feedback signals.

## 20 DETAILED DESCRIPTION OF THE PREFERRED EMBODIMENTS

[28] Figure 1 is a block diagram of an audio system 100 having feedback identification and feedback reduction or elimination techniques. The audio system uses interpolative feedback identification and may adaptively filter multiple feedback signals using one notch filter. The interpolative feedback identification provides for a single estimate of the feedback frequency achieved from more than one sample of a discrete frequency spectrum representation of a feedback signal. The interpolative feedback identification may include utilizing frequency interpolation by generating a second degree or higher polynomial using one or more samples of the discrete frequency spectrum representation. An accurate representation of the actual

frequency of the feedback signal may be determined, for example, by setting a derivative of the polynomial to zero. A filter, such as a notch filter, may be placed in response to the interpolative feedback identification to reduce or eliminate the feedback signal with little or no effect on the audio signal quality provided by the audio system. The adaptive filtering involves configuring a filter, such as a notch filter, to eliminate multiple feedback signals, which allow other filters to reduce or eliminate other feedback signals. The adaptive filtering may also, or in the alternative, reduce processor memory and/or computational requirements of the audio system.

10 [29] The audio system 100 includes an audio input, i.e. a microphone 102, for receiving an audio signal. The microphone 102 is coupled with a microprocessor 104, which is capable of controlling operation of the audio system 100. The microprocessor 104 may perform any analog to digital conversions of audio signals received and digital signal processing. The microprocessor 104 is further capable of performing digital to analog conversions of audio provided by the audio system 100. The microprocessor 104 is coupled with an amplifier 106 capable of amplifying an output audio signal. Amplifier 106 is coupled with a loudspeaker 108 for providing the output audio signal to a user of the audio system. While a particular configuration is shown, the audio system may have other configurations, including those with fewer or additional components.

20 [30] Figure 2 is a flow chart of a method for identifying and reducing and/or removing a feedback signal in an audio system. A time-domain audio signal  $s[n]$  from the microphone 104 is received 200 at microprocessor 104. Audio feedback may result when one or more portions of the audio provided from loudspeaker 108 is received at microphone 102, thereby causing an unstable, closed-loop system. Microprocessor 104 converts 202 the time-domain audio signal into a sampled frequency-domain signal  $|S(k)|$ . Microprocessor 104 may use windowing techniques such as Rectangular, Hamming, Bartlet, and like techniques to compute the frequency domain signal. The microprocessor 104 may then detect 204 the feedback. The detection of feedback may include performing frequency spectrum



analysis such Discrete Fourier Transform (DFT), Spectral Estimation, Filter Banks, and like techniques. Samples of the frequency domains signal may be used in interpolating 206 to determine the frequency of the feedback signal, and the feedback signal may be filtered 208. Interpolating 206 and filtering 208 will be discussed  
 5 further below with respect to Figure 10.

[31] Figures 3-10 illustrate detecting of the feedback signal by microprocessor 104. Figure 3 illustrates a time-domain feedback signal  $s[n]$ . Figure 4 illustrates a frequency domain signal  $|S(e^{j\omega})|$  resulting from converting the feedback signal  $s[n]$  to the frequency domain using, for example, the Discrete Time Fourier Transform (DTFT). Figure 5 illustrates a time-domain window function  $w[n]$ . Figure 6  
 10 illustrates the DTFT ( $|W(e^{j\omega})|$ ) of the window function  $w[n]$ . Figure 7 illustrates the product of the time-domain feedback signal  $s[n]$  with the time-domain window function  $w[n]$ . Figure 8 illustrates the windowed frequency domain signal  $|\hat{S}(e^{j\omega})|$  centered about the frequency domain feedback signal  $|S(e^{j\omega})|$ , resulting from taking  
 15 the DTFT of the product of  $s[n]$  and  $w[n]$ . Figure 9 illustrates the sampled frequency domain signal  $|\hat{S}[k]|$  resulting from taking the DFT of the product of  $s[n]$  and  $w[n]$ . This is, for example, equivalent to sampling the windowed frequency domain feedback signal  $|\hat{S}(e^{j\omega})|$  of Figure 8 at equally spaced frequency intervals. Figure 10 illustrates a portion of the sampled, windowed frequency domain signal  $|S[k]|$  of  
 20 Figure 9, specifically showing a more detailed view around a main lobe of the feedback signal. The frequency spectrum signals illustrated in Figures 4, 6 and 8 are DTFT. The frequency spectrum signals illustrated in Figures 9 and 10 are DFTs. Other frequency spectrum analysis techniques may be utilized in converting the time-domain signal to the frequency domain, and analyzing the frequency domain signal.

25 [32] In the flowchart of Figure 2, the interpolating 206 provides a single representation/estimation of a feedback frequency determined from multiple samples of the discrete frequency spectrum representation of the frequency signal. The interpolative feedback identification may be determined using frequency interpolation techniques, for example, as will be described with respect to the graph

of Figure 10, where each frequency sample defines a frequency bin. The notations used in Figure 10 are as follows:

[33]  $B_{estimate}$  = The estimated frequency of the feedback signal.

[34]  $B_p$  = Peak (maximum) bin number.

5 [35]  $B_{p-1}$  = Bin just below (in frequency) the peak bin number.

[36]  $B_{p+1}$  = Bin just above (in frequency) the peak bin number.

[37]  $A_{estimate}$  = Amplitude at the estimated frequency of the feedback.

[38]  $A_p$  = Amplitude of the peak bin.

[39]  $A_{p-1}$  = Amplitude of the bin just below (in frequency) the peak bin.

10 [40]  $A_{p+1}$  = Amplitude of the bin just above (in frequency) the peak bin.

[41]  $B_{estimate}$  is the estimated frequency of the feedback signal which may be determined using the interpolation techniques described below. Ideally, the frequency  $B_{estimate}$  will coincide with the actual frequency of the feedback signal. In any event, the frequency  $B_{estimate}$  is typically a more accurate estimate of the actual frequency of the feedback signal than the frequency  $B_p$  which is selected by systems of the prior art.

[42] Interpolative feedback identification such as frequency interpolation provides a more accurate estimate of the actual frequency of feedback, and may be determined using samples of the DFT  $|S[k]|$ . Using the samples of the DFT signal  $|S[k]|$ , a unique quadratic (or higher order polynomial) may be generated which resembles the original main lobe of the DTFT representing the feedback signal. A polynomial may be reconstructed from the sample points of the DFT  $|S[k]|$ . An interpolating polynomial for degree N-1 is illustrated as a LaGrange polynomial by:

$$P(x) = \frac{(x-x_2)(x-x_3) \wedge (x-x_N)}{(x_1-x_2)(x_1-x_3) \wedge (x_1-x_N)} y_1 + \frac{(x-x_1)(x-x_3) \wedge (x-x_N)}{(x_2-x_1)(x_2-x_3)(x_2-x_N)} y_2 + \wedge + \frac{(x-x_1)(x-x_2) \wedge (x-x_{N-1})}{(x_N-x_1)(x_N-x_2) \wedge (x_N-x_{N-1})} y_N$$

25 Other interpolating polynomial techniques may be used, including polynomial interpolation, rational function interpolation, cubic spline interpolation and the like.

[43] Applying the LaGrange polynomial equation to frequency interpolation (here, for a 2<sup>nd</sup> order quadratic) yields a feedback frequency equation  $f(B)$  of:

$$f(B) = \frac{(B - B_p)(B - B_{p+1})}{(B_{p-1} - B_p)(B_{p-1} - B_{p+1})} A_{p-1} + \frac{(B - B_{p-1})(B - B_{p+1})}{(B_p - B_{p-1})(B_p - B_{p+1})} A_p + \frac{(B - B_{p-1})(B - B_p)}{(B_{p+1} - B_{p-1})(B_{p+1} - B_p)} A_{p+1}$$

5 [44] A peak of the quadratic curve, and thus an estimate/representation of the frequency of the feedback signal may be determined by solving for a maximum of  $f(B)$ . Solving for the maximum may be accomplished, for example, by taking the derivative of  $f(B)$ , and setting the derivative to zero, yielding the estimated feedback frequency  $B_{\text{estimate}}$  as:

10 [45]

$$B_{\text{estimate}} = \frac{[A_{p-1} * (B_p + B_{p+1})(B_p - B_{p-1})(B_p - B_{p+1})(B_{p+1} - B_{p-1})(B_{p+1} - B_p)]}{2}$$

$$+ \frac{[A_p * (B_{p-1} + B_{p+1})(B_{p-1} - B_p)(B_{p-1} - B_{p+1})(B_{p+1} - B_{p-1})(B_{p+1} - B_p)]}{2}$$

15  $+ \frac{[A_{p+1} * (B_{p-1} + B_p)(B_{p-1} - B_p)(B_{p-1} - B_{p+1})(B_p - B_{p-1})(B_p - B_{p+1})]}{2}$

[46] The pole of the quadratic curve provides a more accurate representation of the frequency of the feedback signal than the frequency  $B_p$  of the peak bin alone. Where it is known that, prior to the interpolation,  $A_p$  is greater than both  $A_{p+1}$  and  $A_{p-1}$ , it may be determined that the interpolated polynomial has no minimum at this location, and only a maximum. Thus, taking the derivative of the interpolation polynomial and setting it to zero yields the maximum, and thus the more accurate representation of the frequency of the feedback signal than the frequency  $B_p$ . However, where it is not known prior to the interpolation that  $A_p$  is greater than both

20

$A_{p+1}$  and  $A_{p-1}$ , it may be necessary to determine that the frequency at  $B_{\text{estimate}}$  is a maximum and not a minimum of the quadratic equation.

[47] To determine that the frequency at  $B_{\text{estimate}}$  is a maximum (and not a minimum) of the quadratic equation, a value  $A_{\text{estimate}}$  may be computed by the microprocessor 104 using the equation for  $f(B)$  above, representing the amplitude of the feedback signal at the interpolated frequency  $B_{\text{estimate}}$ .  $A_{\text{estimate}}$  may be compared with the values  $A_{p+1}$  and  $A_{p-1}$ , which are amplitudes of the feedback signal at corresponding frequencies  $B_p$  and  $B_{p+1}$ , to ensure that  $A_{\text{estimate}}$  has the highest amplitude.

[48] The interpolating 206 of Figure 2 provides a more accurate estimate of the actual frequency of feedback signal. Using the frequency estimate  $B_{\text{estimate}}$ , a filter may be configured for filtering 208 the feedback of the audio signal. The filter may be a bandwidth notch filter. Other filters may be used. Since a close estimate for the frequency of the feedback signal has been identified using frequency interpolation, the bandwidth notch filter may be configured (i.e., coefficients calculated therefore including Quality Factor and/or gain/cut-depth) by the microprocessor 104 as a narrow bandwidth notch filter capable of filtering-out the frequency of the feedback signal. The microprocessor 104 may also minimize at least one of a bandwidth and a cut-depth of the notch filter. The configured filter may then be placed at the frequency  $B_{\text{estimate}}$  (i.e. designed with a center frequency of  $B_{\text{estimate}}$ ). Such filtering may be employed utilizing filtering techniques such as Finite Impulse Response (FIR) and Infinite Impulse Response (IIR) techniques, or any other filtering technique sufficient for filtering out the feedback signal as would be appreciated by one skilled in the art. Thus, identifying the frequency of the feedback signal using interpolative feedback identification allows for more accurate placement of the notch filter at the frequency of the feedback signal, and thus is more accurately configured for filtering-out the feedback signal.

[49] Figure 10 illustrates an example of interpolation by generating a polynomial which models the original main lobe of the frequency spectrum, where the interpolation is carried-out by solving for a maximum of the polynomial by

derivation. One skilled in the art would realize that any interpolation techniques may be utilized to identify the feedback frequency. For example, additional frequency bins may be interspaced between samples of the sample frequency domain signal shown in Figure 10, each interspaced bin having zero energy value. The sampled  
5 frequency domain signal may then be passed through a low pass filter resulting in an interpolated sampled spectrum. Using the interpolated sampled spectrum, one could identify a maximum of the filtered frequency spectrum to obtain a more accurate estimate of the feedback signal frequency.

[50] Figures 11 and 12 illustrate graphs comparing characteristics of prior art  
10 notch filters with notch filters configured in accordance with interpolative feedback identification. The sampled frequency bin having a maximum amplitude  $B_p$  in Figure 10, may correspond to 994 Hz in Figures 11 and 12. A more accurate representation of the frequency of the feedback signal,  $B_{estimate}$  in Figure 10, may correspond to 1000 Hz in Figures 11 and 12. The sampled frequency bins and  
15 frequency of the feedback signal may have other frequencies. As shown at Figures 11 and 12, prior art feedback identification techniques result in a notch filter being configured to filter out frequencies at the maximum bin frequency 994 Hz, and thus must have an increased bandwidth as shown by line 1100 Figure 11, or increased cut-depth as shown by line 1200 of Figure 12, to ensure that the gain (G) of  
20 the filter at the actual frequency of the feedback is sufficient for filtering the feedback signal.

[51] In contrast, feedback identification techniques using interpolative  
feedback identification provide a more accurate representation (here about 1000 Hz)  
of the actual frequency of feedback. Accordingly, a notch filter having  
25 characteristics shown at 1105 and 1205 of Figures 11 and 12 may be placed at the more accurate estimate for the actual frequency of the feedback signal. Because the filter is more accurately placed, it may be more narrowly tailored (i.e. reduced bandwidth and/or cut-depth) while ensuring that the gain is sufficient at the  
frequency of the feedback signal to eliminate or reduce the feedback signal, and  
30 having little or no effect on the quality of the signal provided at the loudspeaker 108,

or in any event, less of an effect on the audio quality than notch filters configured using prior art feedback identification techniques.

[52] Figure 13 is a flow chart of a method for providing adaptive filtering of feedback in an audio system. Frequencies of a plurality of feedback signals are identified/estimated 1300 by the microprocessor 104. Such frequencies may be identified as described above using interpolative feedback identification, or in any other fashion. The microprocessor 104 determines 1302 whether the frequencies of feedback signals are within a frequency window covering a specified frequency range. The frequency range covered by the frequency window may be predetermined and/or configurable, and may vary depending on the frequency band being examined. The specified frequency range covered by the frequency window will be discussed further below with respect to Figures 14 and 15.

[53] The microprocessor 104 filters 1304 the feedback signal within the frequency range covered by the frequency window. The microprocessor 104 configures a filter for filtering out any frequencies a feedback signal determines to be within the frequency range. The filter may be a notch filter or other type of filter. The microprocessor may determine filter coefficients such as quality factor, cut-depth and a center frequency for the filter.

[54] Figure 14 is a graph illustrating a frequency window covering a specified frequency range for time-domain representations of feedback signals, which may be utilized in providing the adaptive filtering discussed above with respect to Figure 13. As shown in Figure 14, a frequency window represented generally at 1405 may cover a specified frequency range, for example,  $\alpha f$ . Where two feedback frequencies, for example feedback frequency  $f_1$  and feedback frequency  $f_2$  lie within the frequency window 1405, it may be determined 1302 that adaptive filtering will be utilized to configure a single filter to filter out the feedback frequencies.

[55] To determine whether the feedback frequencies lie within the frequency window 1405, a frequency differential  $\Delta f$  may be determined between feedback frequencies, for example by subtracting one frequency from another. For example, as shown in Figure 14,  $\Delta f$  may be determined by subtracting the frequency  $f_1$

representing a first frequency at which feedback is located from  $f_2$  representing a second frequency at which feedback is located. Where the value  $\Delta f$  is less than  $\alpha f$ , and thus the frequency range covered by the frequency window 1405, it may be determined that the feedback located at frequencies  $f_1$  and  $f_2$  may be adaptively filtered by a single filter.

[56] A filter may be configured, for example by the microprocessor 104 at a center frequency  $f_c$  within the frequency window 1405 having sufficient quality factor and/or cut-depth to filter out the feedback at the frequencies  $f_1$  and  $f_2$ .

[57] Concurrently or subsequently, if a feedback signal is identified as being located at a frequency  $f_3$ , for example as shown in Figure 14, the microprocessor 104 may determine whether the frequency differential  $\Delta f$  between  $f_3$  and  $f_c$  is less than the frequency range  $\alpha f$  covered by the frequency window 1405. Where it is determined that the newly calculated  $\Delta f$  is less than  $\alpha f$ , the microprocessor 104 may determine that the feedback identified at  $f_3$  may be adaptively filtered utilizing the filter at  $f_c$ , and thus reconfigure the filter centered at  $f_c$  (i.e., reconfigure the quality factor, cut-depth and/or  $f_c$ ) to filter out the feedback identified at the frequencies  $f_1$ ,  $f_2$  and  $f_3$ .

[58] Alternatively, instead of determining the frequency differential between  $f_3$  and  $f_c$ , the microprocessor 104 may instead determine a frequency differential  $\Delta f$  between  $f_3$  and  $f_1$  for comparing with the frequency range  $\alpha f$  of the frequency window 1405 in determining whether the feedback frequencies  $f_1$ ,  $f_2$  and  $f_3$  may be adaptively filtered by a single filter. As additional feedback frequencies are concurrently and/or subsequently identified, the microprocessor 104 may determine whether to employ additional filters, or to utilize existing filters to cover the concurrently or subsequently identified frequencies of feedback.

[59] In addition, the microprocessor 104 may further utilize algorithms that may minimize the number of filters necessary to filter out the identified feedback frequencies. In Figure 14, the frequency of the feedback frequency  $f_1$  may be 1000Hz, where the feedback frequency  $f_2$  may be 1012 Hz and the feedback frequency  $f_3$  may be 1024 Hz. The specified frequency range  $\alpha f$  of the frequency window 1405 may be any value, for example, 6 Hz, 12 Hz, 20 Hz, 100 Hz or any other value. The specified

frequency range  $\alpha f$  may vary across the frequency spectrum, as a function of the frequency of the particular feedback frequencies being examined. For example, the frequency range  $\alpha f$  may increase logarithmically as the particular frequency being examined for feedback increases. Thus, at lower frequencies,  $\alpha f$  may have a smaller value than  $\alpha f$  at higher frequencies. In addition, the value of  $\alpha f$  defining the frequency window 1405 may be configurable by a user of the system 100.

[60] The graph of Figure 14 describes how the determining 1302 may be accomplished for feedback signals represented in the time-domain. The determining 1310 may similarly be carried-out for identified feedback signals in the frequency domain, for example as described with respect to the graph of Figure 15.

[61] Figure 15 is a graph illustrating a frequency window covering a specified frequency range for frequency domain representations of feedback signals, which may be utilized for the adaptive filtration discussed above. A frequency window 1505 is shown, covering a specified frequency range represented by a particular number of frequency bins (i.e., frequency samples)  $\alpha B$ . To determine 1302 whether the feedback frequencies lie within the frequency window 1505, a frequency differential represented here as a number of frequency bins,  $\Delta B$ , may be determined between feedback frequency bins, for example by subtracting one feedback frequency bin from another. As shown in Figure 15,  $\Delta B$  may be determined by subtracting the frequency bin# B328 representing a first frequency at which feedback is located from the frequency bin# B326 representing a second frequency at which feedback is located. Where the value  $\Delta B$  is less than  $\alpha B$ , and thus the frequency range covered by the frequency window 1505, it may be determined that the feedback located at frequency bins B328 and B326 may be adaptively filtered by a single filter.

[62] A filter may be configured, for example by the microprocessor 104 at a center frequency  $f_c$  within the frequency window 1505 having sufficient quality factor and/or cut-depth to filter out the feedback at the frequency bins B326 and B328.

[63] Concurrently or subsequently, if a feedback signal is identified as being located at a frequency bin #B333, for example as shown in Figure 15, the



microprocessor 104 may determine whether the frequency differential  $\Delta B$  between the frequency bin #B333 and  $f_c$  is less than the specified frequency range  $\alpha B$  covered by the frequency window 1505. Where it is determined that the newly calculated  $\Delta B$  is less than  $\alpha B$ , the microprocessor 104 may determine that the feedback identified at  
5 frequency bin #B333 may be adaptively filtered utilizing the filter at  $f_c$ . The microprocessor 104 may reconfigure the filter centered at a center frequency  $f_c$  (i.e., reconfigure the quality factor, cut-depth and/or  $f_c$ ) to filter out the feedback identified at the frequencies represented by frequency bins 326, 328 and 333. In Figure 15, the center frequency  $f_c$  is shown, by example, at bin #B327.

10 [64] Similar to as discussed above with respect to Figure 14, instead of determining the frequency differential between bin #B333 and  $f_c$ , the microprocessor 104 may instead determine a frequency differential  $\Delta B$  between bins B333 and B326. This frequency differential  $\Delta B$  may be compared with the frequency range  $\alpha B$  of the frequency window 1505 to determine whether the feedback frequencies represented at  
15 bins B326, B328 and B333 may be adaptively filtered by a single filter. As additional feedback frequencies are concurrently and/or subsequently identified, the microprocessor 104 may determine whether to employ additional filters, or to utilize existing filters to cover the concurrently or subsequently identified frequencies of feedback.

20 [65] Additionally, and as discussed above, the microprocessor 104 may further utilize algorithms that may minimize the number of filters necessary to filter out the identified feedback frequencies. The specified frequency range  $\alpha B$  of the frequency window 1505 is shown in Figure 15 as being 3 frequency bins, where the bin #326 may represent a frequency sample at 1000Hz, and spacing between frequency samples/bins  
25 may be approximately 6 Hz. However, similar to as discussed above with respect to Figure 14, it will be appreciated by one skilled in the art that  $\alpha B$  may be any number of frequency bins, for example 2, 3, 5 or 10 frequency bins, and that the frequency differential represented by  $\alpha B$  may vary as a function of the feedback frequencies being

examined. In addition, the value of  $\alpha B$  defining the frequency window 1505 may be configurable by a user of the system 100.

5 [66] Figure 16 illustrates a graph showing characteristics of adjacently placed notch filters that may benefit from the adaptive filtering discussed herein. Feedback has been identified at frequencies of  $f_1$  equal to about 1000 and  $f_2$  equal to about 1012 Hz. To eliminate the feedback identified at these frequencies, notch filters may be utilized having the characteristics 1600 and 1605. The characteristics 1600 include a Quality Factor equal to about 128 and a cut-depth equal to about -6dB to eliminate or reduce the feedback. The characteristics 1605 include a Quality Factor equal to about 128 and a cut-depth equal to about -6dB to eliminate or reduce the feedback. However, in utilizing adaptive filtering, microprocessor 104 is capable of determining that the frequency differential  $\Delta f$  between feedback frequencies at frequencies  $f_1$  and  $f_2$  are within a frequency range  $\alpha f$  defining a frequency window, where  $\alpha f$  may be 15 Hz. Microprocessor 104 may configure a single notch filter to filter out the feedback from both identified feedback frequencies.

15 [67] In Figure 17, characteristics of a notch filter configured by the microprocessor 104 is shown at 1700. The characteristics indicate a notch filter designed for a center frequency  $f_c$  of about 1006 Hz and having a Quality Factor of equal to about 45, and a cut-depth equal to about -6dB. The notch filter is placed between the two identified frequencies, here  $f_1$  at about 1000 Hz and  $f_2$  at about 1012 Hz, to filter out the feedback signal frequencies. The notch filter may be placed (i.e. designed with a center frequency) at a midpoint of the frequencies of identified feedback, here about 1006 Hz. The notch filter may be placed at any other frequency between the identified feedback frequencies, or within the frequency window being examined (not shown), sufficient for filtering out the identified feedback. Where more than two frequencies of feedback signals are determined to fall within the frequency range  $\alpha f$ , an average frequency may be calculated for the determined frequencies of feedback, where the filter is placed at the average frequency. Alternatively, a midpoint frequency between the greatest and lowest frequencies

determined to be within the frequency range  $\alpha f$  defining the frequency window may be selected for placement of the notch filter.

[68] Thus, instead of requiring two or more notch filters to filter out multiple feedback signals within the frequency window defined by the frequency range  $\alpha f$ , a single notch filter may be utilized. Hence, the other notch filter(s) available in the audio system may be used to eliminate or reduce feedback at other frequencies. Rather than having additional notch filters, reducing the number of notch filters for filtering feedback signals may reduce the memory and/or processing requirements of microprocessor 104. The filtering may be accomplished as software executed on the microprocessor 104.

[69] Further, multiple sets of frequencies of feedback signals may be identified by the microprocessor 104, where the microprocessor 104 configures a notch filter to filter the feedback signals corresponding to each set of feedback frequencies.

[70] The audio system 100 discussed above may be utilized in cellular telephones, public address systems, speakerphones having duplex operation, or any other audio system that may suffer from feedback. The microphone 102 may be any input transducer sufficient for receiving audio into the audio system 100. The microprocessor 104 may be any microprocessor capable of performing the functionality/processing, including converting time-domain signals to sampled frequency domain signals. Further, although not shown, the microprocessor 104 may include, or may be coupled with, an external storage media such as computer memory that may include computer programming, executable on the microprocessor 104, for carrying out one or more of the functionalities described herein. The storage medium may be magnetic, optical or any other storage media capable of providing programming for the microprocessor 104.

[71] The loudspeaker 108 may be any speaker capable of providing the output audio from the audio system 100. Alternatively, hardware components not shown may be coupled with the microprocessor 104 for performing the sampled frequency domain conversion where the microprocessor 104 does not possess such functionality. The filtering may be accomplished using software, hardware or a

combination, and need not be limited to notch filtering techniques. The software may be executable on a microprocessor such as performing digital signal processing or the like. The hardware may be coupled with the microprocessor 104, which may configure the hardware to achieve desired processing and/or filtering characteristics.

5 [72] In addition, the values illustrated and discussed in relation to the Figures are exemplary, and are not limitations on the feedback identification and elimination or reduction system. Further, the value for the frequency range  $\alpha f$  with respect to adaptive filtering may be any value while achieving at least some of the advantages discussed herein. The frequency range  $\alpha f/\alpha B$  may be increased (made larger) to  
10 reduce the number of filters required to eliminate feedback. A lower number of filters may be desired where the number of feedback signals outnumber the number of filters available for filtering feedback, or where a processor performing the filtering has limited memory and/or processing capabilities. The frequency window defined by the frequency range  $\alpha f/\alpha B$  may be determined based on considerations  
15 within the particular audio system utilized, and may be user-configurable. Such considerations may include selection of a frequency range which allows frequencies of feedback signals to be combined without unduly affecting the audio quality provided by the audio system. However, different audio systems have varying requirements as to the audio quality provided thereby. For example, a public address  
20 system may have less stringent audio quality requirements than an audio system that may be used in a concert hall or the like. A larger frequency range value  $\alpha f/\alpha B$  may be desired for the former than for the latter to account for desired audio quality.

[73] Further, one skilled would realize that various techniques may be employed in identifying which frequencies of feedback within the frequency range  
25  $\alpha f/\alpha B$ . Further, the microprocessor may utilize various techniques in grouping identified feedback signal sets which are each to be filtered by a single filter, where the technique may minimize the number of filters required for filtering the identified feedback signals.

[74] The audio system 100 may perform both interpolative feedback  
30 identification in identifying frequencies of feedback signals, and adaptive filtration

for configuring a filter-to-filter out multiple frequencies of feedback signals. The audio system 100 need not perform the feedback identification using interpolative feedback identification and/or the adaptive filtering. Rather, the audio system 100 may be utilized in identifying the frequencies of feedback using interpolative  
5 feedback identification while being coupled with additional hardware or microprocessing capabilities which are utilized in eliminating or reducing the identified frequencies of feedback. The hardware may include adaptive filtering. Further, the audio system 100 may perform adaptive filtering using frequencies of feedback identified by external hardware or a processing functionality (which may or  
10 may not include feedback frequencies identified using the interpolative feedback identification).

[75] The illustrations have been discussed with reference to functional blocks identified as modules and components that are not intended to represent discrete structures and may be combined or further sub-divided. In addition, while various  
15 embodiments of the invention have been described, it will be apparent to those of ordinary skill in the art that other embodiments and implementations are possible that are within the scope of this invention. Accordingly, the invention is not restricted except in light of the attached claims and their equivalents.

## CLAIMS

What is claimed is:

1. A method for processing an audio feedback signal, comprising:  
receiving an audio signal;  
5 applying an interpolative feedback identification to a feedback signal within the audio signal; and  
identifying a frequency of the feedback signal in response to the interpolative feedback identification.
- 10 2. The method of Claim 1, where the audio signal is a time-domain audio signal, and further comprising:  
windowing the time-domain signal;  
converting the time-domain signal to a frequency domain signal; and  
applying the interpolative feedback identification to identify the frequency of  
15 the feedback signal between samples of the sampled frequency domain signal.
3. The method of Claim 2, further comprising generating a polynomial responsive to the samples of the frequency domain signal, and applying interpolative feedback identification responsive to the polynomial.  
20
4. The method of Claim 3, further comprising solving for a maximum of the polynomial to identify the feedback frequency.
5. The method of Claim 4, further comprising verifying that the identified  
25 frequency has a greater magnitude than a magnitude of the frequency samples of the windowed frequency domain signal.
6. The method of Claim 2, further comprising:  
calculating coefficients for a notch filter responsive to the identified frequency  
30 of the feedback signal; and  
filtering out the identified frequency of the feedback signal using the notch filter.

7. The method of Claim 6, further comprising minimizing at least one of a bandwidth and a cut-depth of the notch filter responsive to the identified frequency of the feedback signal.

5

8. The method of Claim 1, further comprising filtering the audio signal to eliminate the identified feedback responsive to the interpolative feedback identification.

10 9. The method system of Claim 1, where there are a plurality of feedback signals within the audio signal, and further comprising identifying a plurality of frequencies corresponding to the plurality of feedback signals responsive to the interpolative feedback identification.

15 10. The method of Claim 9, further comprising filtering the audio signal to eliminate the plurality of frequencies of identified feedback responsive to the interpolative feedback identification.

20 11. The method of Claim 9, further comprising adaptively filtering at least two adjacent identified frequencies of feedback signals responsive to the interpolative feedback identification with a single filter by configuring the filter to filter the at least two adjacent identified frequencies.

25 12. The method of Claim 11, where the filter is a notch filter, and further comprising determining coefficients for the notch filter to minimize at least one of a bandwidth and a cut-depth of the notch filter responsive to the identified frequencies of the feedback signal.

30 13. The method of Claim 11, further comprising selecting the at least two adjacent identified frequencies of feedback within a specified frequency range.

14. The method of Claim 13, where the specified frequency range is variable responsive to a feedback frequency being examined.
15. The method of Claim 11, further comprising:  
5 determining a minimum identified frequency and a maximum identified frequency which will be filtered;  
selecting an interior frequency between the minimum identified frequency and the maximum identified frequency; and  
10 configuring the filter at the selected interior frequency.
16. An audio system comprising:  
an audio signal port for receiving an audio signal;  
a processor coupled with the signal port for applying interpolative feedback identification to a feedback signal within the audio signal, and identifying a frequency  
15 of a feedback signal within the audio signal responsive to the interpolative feedback identification.
17. The audio system of Claim 16, where the audio signal is a time-domain audio signal, and further comprising:  
20 the processor being capable of windowing the feedback signal, converting the windowed feedback signal to a frequency domain signal, and identifying the frequency of the feedback signal between samples of the sampled frequency domain signal responsive to the interpolative feedback identification.
- 25 18. The audio system of Claim 17, where the processor identifies the frequency of the feedback signal between samples using the interpolative feedback identification by generating a polynomial responsive to the samples of the frequency domain signal.
- 30 19. The audio system of Claim 18, where the processor identifies the frequency of the feedback signal responsive to the interpolative feedback identification by solving for a maximum of the generated polynomial.



20. The audio system of Claim 19, where the processor identifies the frequency of the feedback signal responsive to the interpolative feedback identification by verifying that the identified frequency has a greater energy magnitude than an energy magnitude of the frequency samples of the frequency domain signal.

5

21. The audio system of Claim 17, further comprising the processor calculating coefficients for a notch filter responsive to the identified frequency of the feedback signal.

10 22. The audio system of Claim 21, where the processor calculates coefficients for the notch filter by minimizing at least one of a bandwidth and a cut-depth of the notch filter responsive to the identified frequency of the feedback signal.

15 23. The audio system of Claim 16, further comprising a filter coupled with the processor and the audio signal port, for filtering the audio signal to reduce the feedback at the identified feedback frequency responsive to the interpolative feedback identification.

20 24. The audio system of Claim 23, where the filter is a notch filter tuned to the identified frequency of the feedback signal.

25 25. The audio system of Claim 16, where there are a plurality of feedback signals within the audio signal, and the processor applies interpolative feedback identification to the plurality of feedback signals, and identifies a plurality of frequencies corresponding to the plurality of feedback signals responsive to the interpolative feedback identification.

30 26. The audio system of Claim 25, further comprising a plurality of filters coupled with the processor and the audio signal port, for filtering the audio signal to eliminate the identified feedback responsive to the interpolative audio.

27. The audio system of Claim 25, further comprising a filter coupled with the processor and the audio signal port, where the processor adaptively filters at least two adjacent identified frequencies of feedback signals by configuring a filter to filter the at least two adjacent identified frequencies, responsive to the interpolative audio.
- 5
28. The audio system of Claim 27, where the filter is a notch filter, and the processor configures the notch filter by determining coefficients for the notch filter to minimize at least one of a bandwidth and a cut-depth of the notch filter responsive to the identified frequencies of the feedback signal.
- 10
29. The audio system of Claim 27, where the processor adaptively filters the at least two adjacent identified frequencies by selecting the at least two adjacent identified frequencies of feedback within a specified frequency range.
- 15
30. The audio system of Claim 29, where the specified frequency range is variable responsive to a feedback frequency being examined.
- 20
31. The audio system of Claim 27, where the processor adaptively filters the at least two adjacent identified frequencies of feedback signals by determining a minimum identified frequency and a maximum identified frequency which will be filtered by the filter, selecting an interior frequency between the minimum identified frequency and the maximum identified frequency, and configuring the filter for filtering at the selected interior frequency.
- 25
32. The audio system of Claim 27, where the processor includes a storage media programmed for applying the interpolative feedback identification to the feedback signal and identifying the frequency of the feedback signal responsive to the interpolative feedback identification.

33. A method of processing audio feedback, comprising:  
receiving an audio signal including multiple feedback signals;  
identifying a plurality of feedback frequencies, each feedback frequency  
corresponding to one of the feedback signals;  
5 determining whether at least two feedback frequencies of the plurality of  
feedback frequencies lie within a specified frequency range; and  
configuring a filter responsive to the determination to filter out the at least two  
determined feedback frequencies.
- 10 34. The method of Claim 33, further comprising configuring a notch filter to filter  
out the at least two determined frequencies.
35. The method of Claim 33, further comprising determining at least one of a  
bandwidth, cut-depth and center frequency of the filter.
- 15 36. The method of Claim 35, further comprising selecting a center frequency for  
the filter which is an average of the at least two feedback frequencies.
37. The method of Claim 33 further comprising determining whether at least two  
20 adjacent feedback frequencies lie within the specified frequency range.
38. The method of Claim 33, further comprising determining whether only two  
feedback frequencies lie within the specified frequency range.
- 25 39. The method of Claim 33 where the specified frequency range is variable  
responsive to a feedback frequency being examined.
40. The method of Claim 33, further comprising applying interpolative feedback  
identification to at least one of the feedback signals, and identifying at least one of the  
30 feedback frequencies responsive to the interpolative feedback identification.

41. A storage media for use on a processor of an audio system, comprising:  
a memory portion programmed for allowing receipt of an audio signal,  
applying an interpolative feedback identification to a feedback signal within the audio  
signal, and identifying a frequency of the feedback signal responsive to the  
5 interpolative feedback identification.
42. The storage media of Claim 41, where the audio signal is a time-domain audio  
signal, and further comprising the memory portion being programmed for:  
windowing the time-domain signal;  
10 converting the time-domain signal to a frequency domain signal; and  
identifying the frequency of the feedback signal between samples of the  
sampled frequency domain signal responsive to the interpolative feedback  
identification.
43. The storage media of Claim 42, further comprising the memory portion being  
15 programmed for generating a polynomial responsive to the samples of the frequency  
domain signal.
44. The storage media of Claim 43, further comprising the memory portion being  
20 programmed for identifying the frequency of the feedback signal by solving for a  
maximum of the polynomial.
45. The storage media of Claim 41, further comprising the memory portion being  
25 programmed for filtering the identified frequency of the feedback signal by  
calculating coefficients for the filter responsive to the identified frequency of the  
feedback signal.
46. The storage media of Claim 45 where the filtering approximates a notch filter,  
and further comprising the memory portion being programmed for minimizing at least  
30 one of a bandwidth and a cut-depth of the approximated notch filter responsive to the  
identified frequency of the feedback signal.

47. A storage media for use on a processor of an audio system, comprising:  
a memory portion programmed for receiving an audio signal including  
multiple feedback frequencies,  
identifying a plurality of feedback frequencies, each feedback frequency  
5 corresponding to one of the feedback signal,  
determining whether at least two feedback frequencies of the plurality of  
feedback frequencies lie within a specified frequency range, and  
configuring a filter to filter out the at least two determined feedback  
frequencies responsive to the determination.

10

48. The storage media of Claim 47, further comprising the memory portion being  
programmed for configuring a notch filter to filter out the at least two determined  
frequencies.

15 49. The storage media of Claim 48, where the notch filter is an approximated  
notch filter, and further comprising the memory portion being programmed for  
approximating the notch filter.

20 50. The storage media of Claim 47, where the memory portion is programmed for  
applying interpolative feedback identification on at least one feedback signal, and  
identifying at least one of the feedback frequencies responsive to the interpolative  
feedback identification.

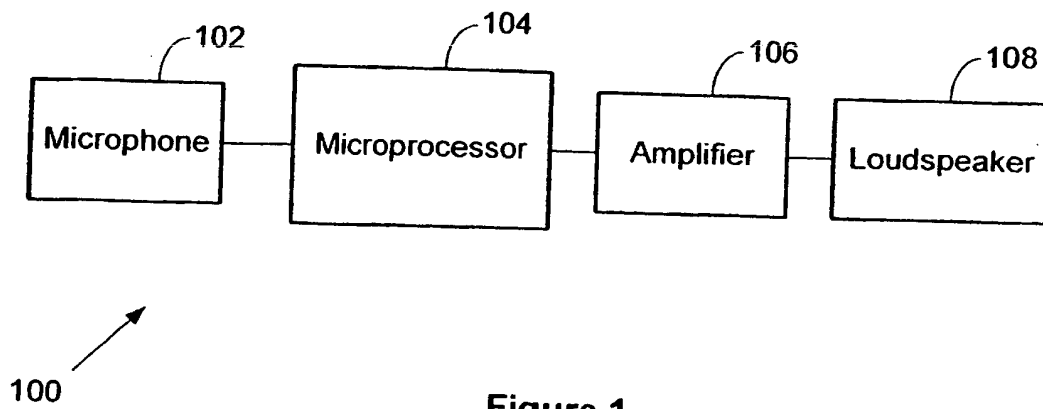


Figure 1

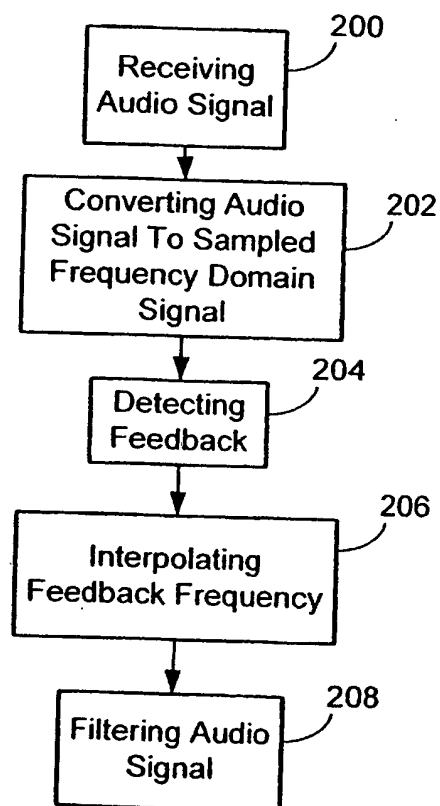


Figure 2

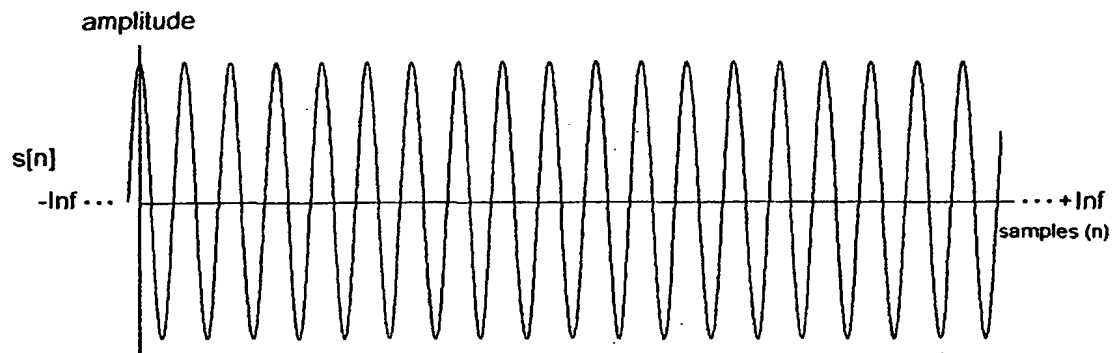


Figure 3

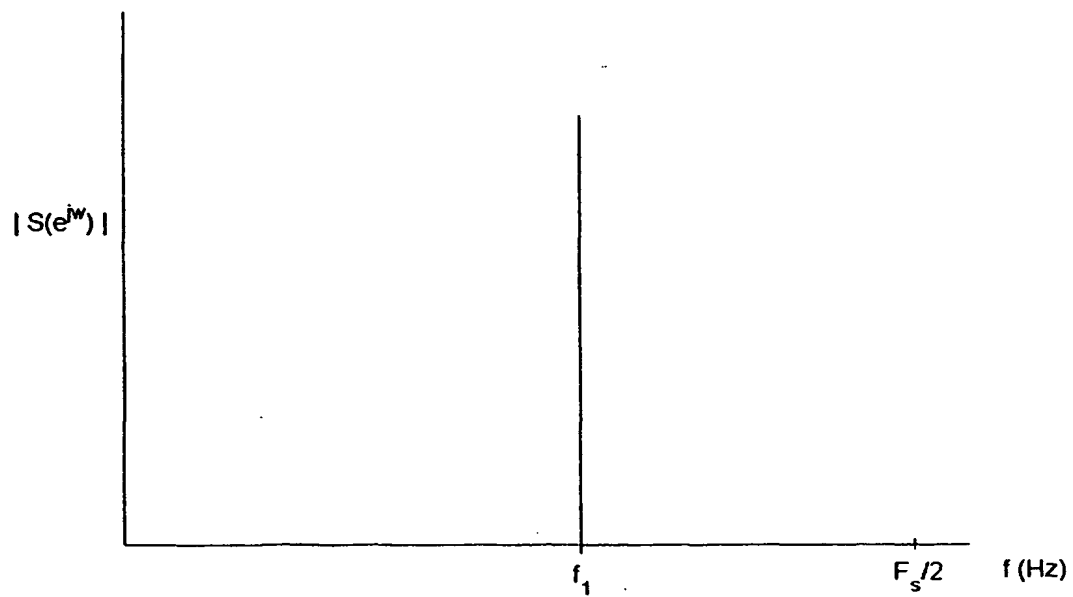


Figure 4

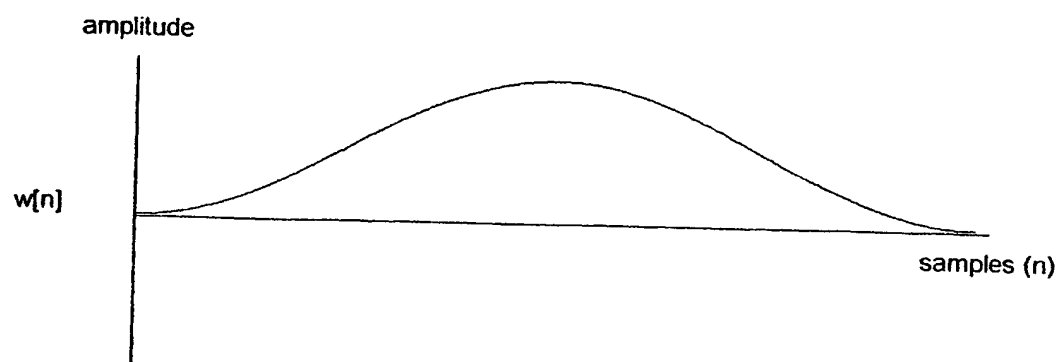


Figure 5

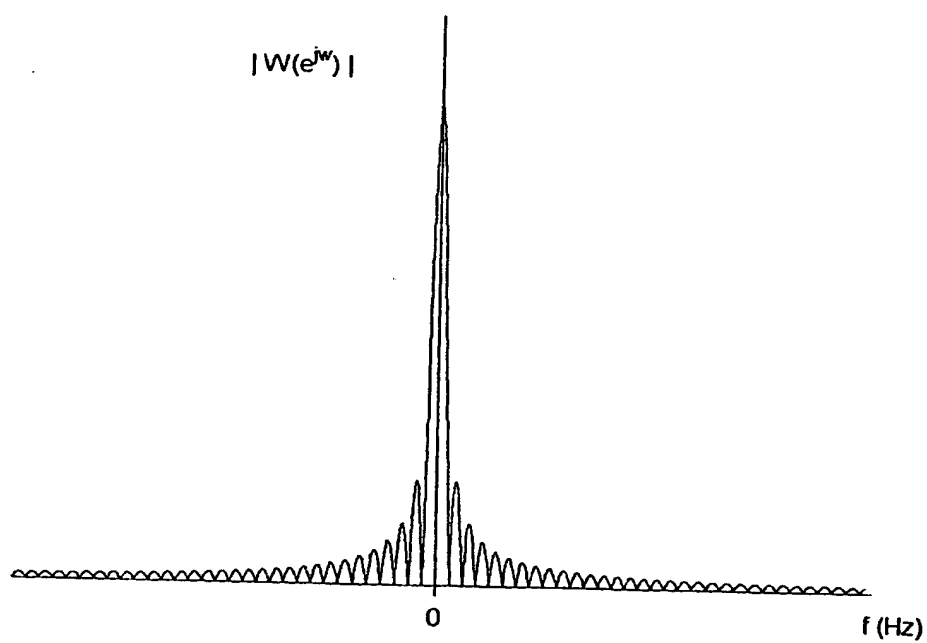


Figure 6



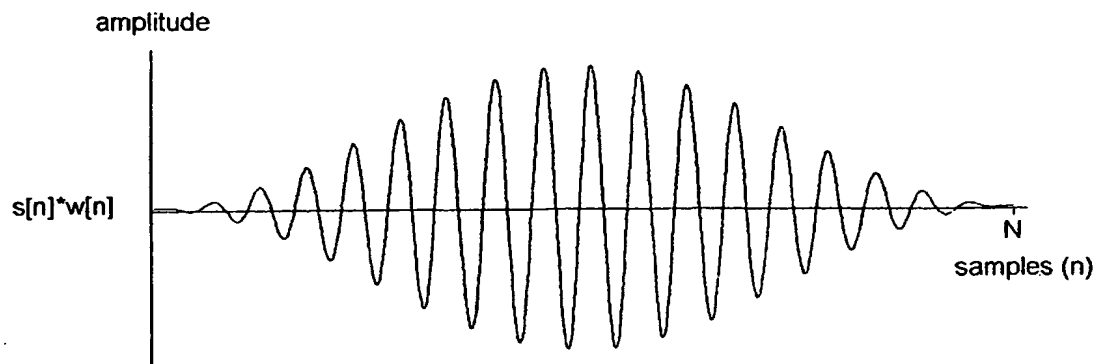


Figure 7

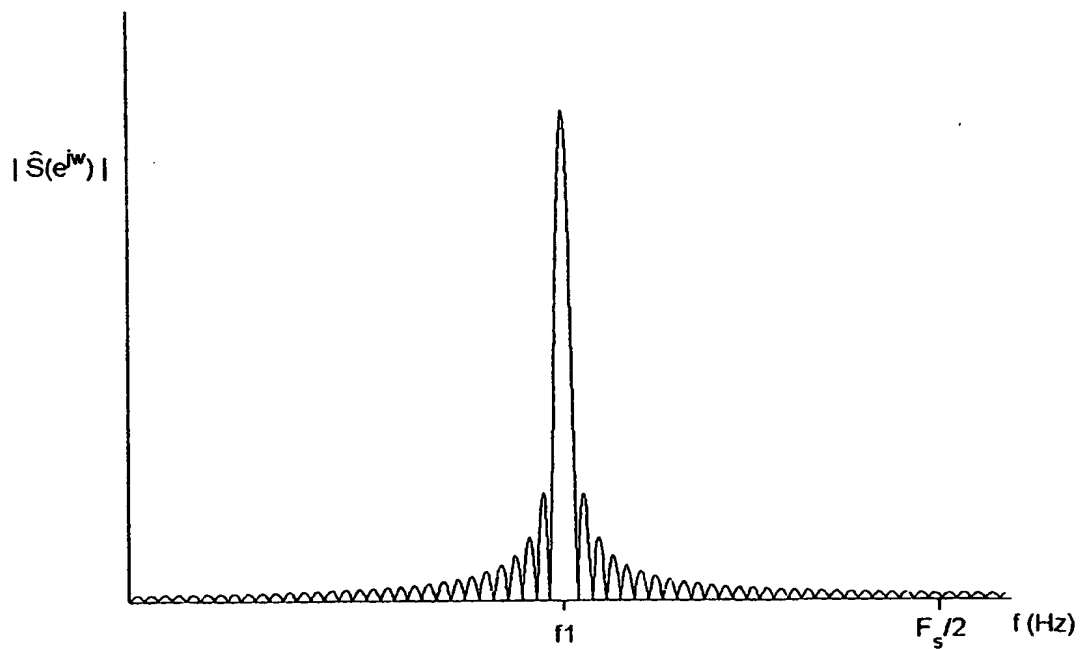


Figure 8

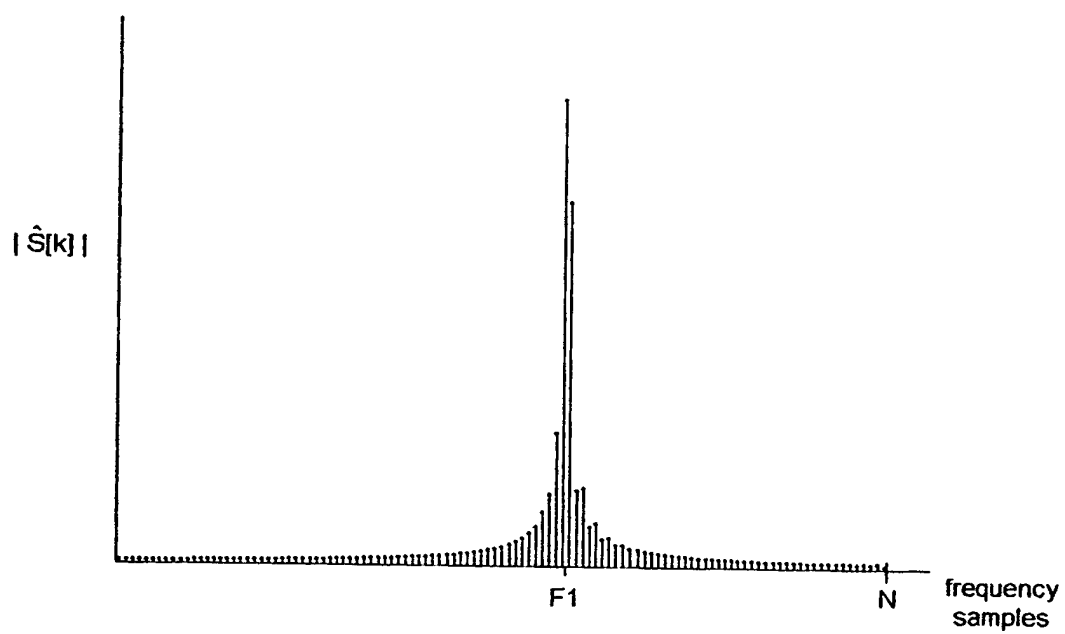


Figure 9

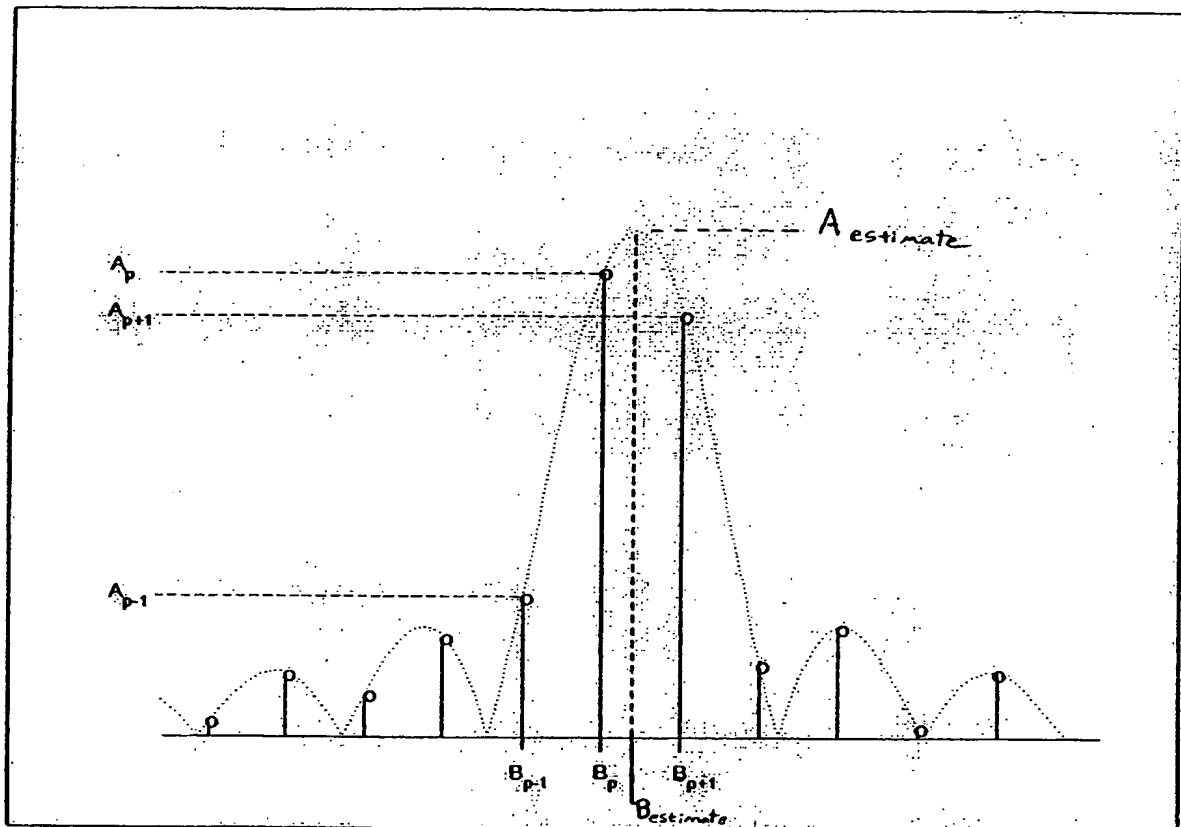


FIGURE 10

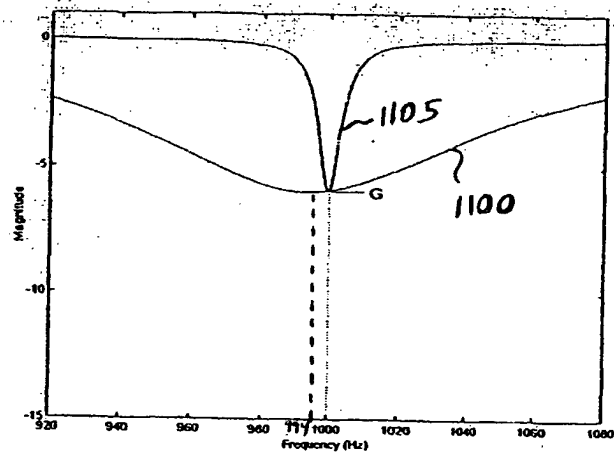


FIGURE 11

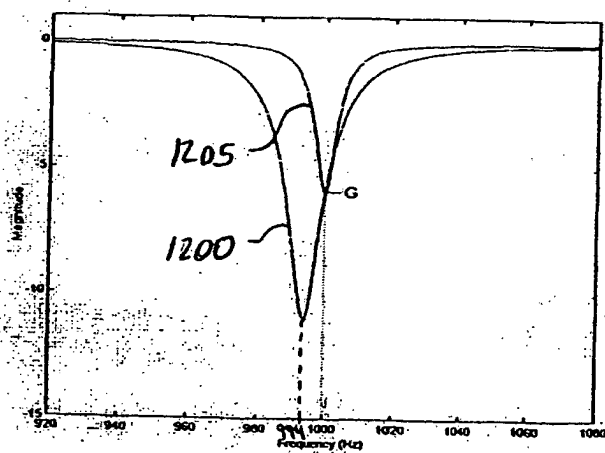
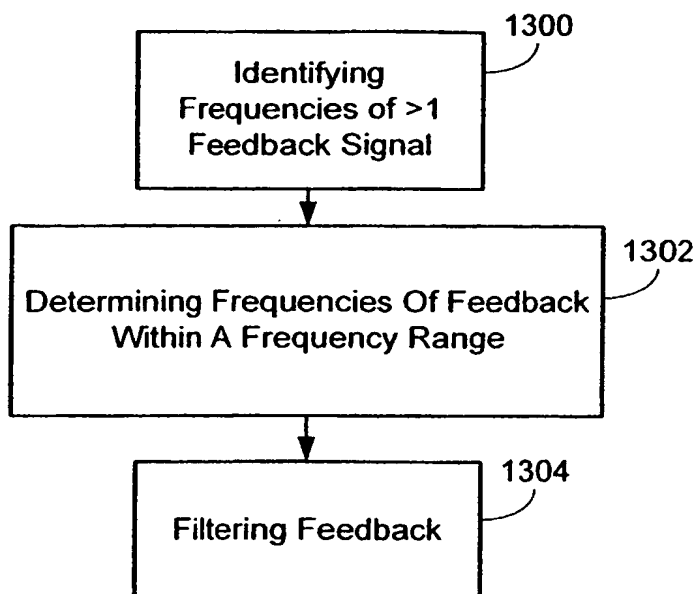


FIGURE 12

**Figure 13**

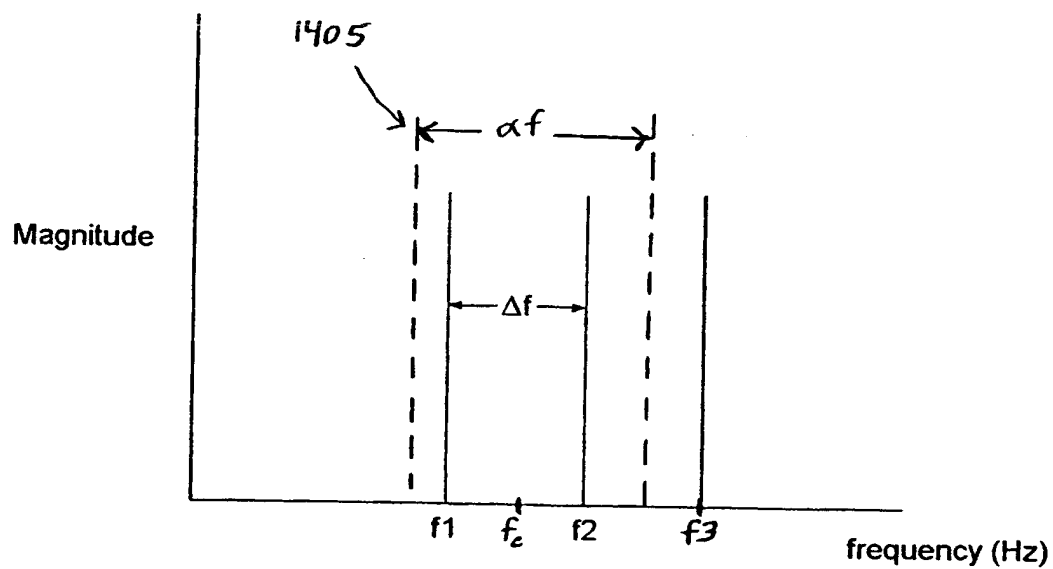


Figure 14

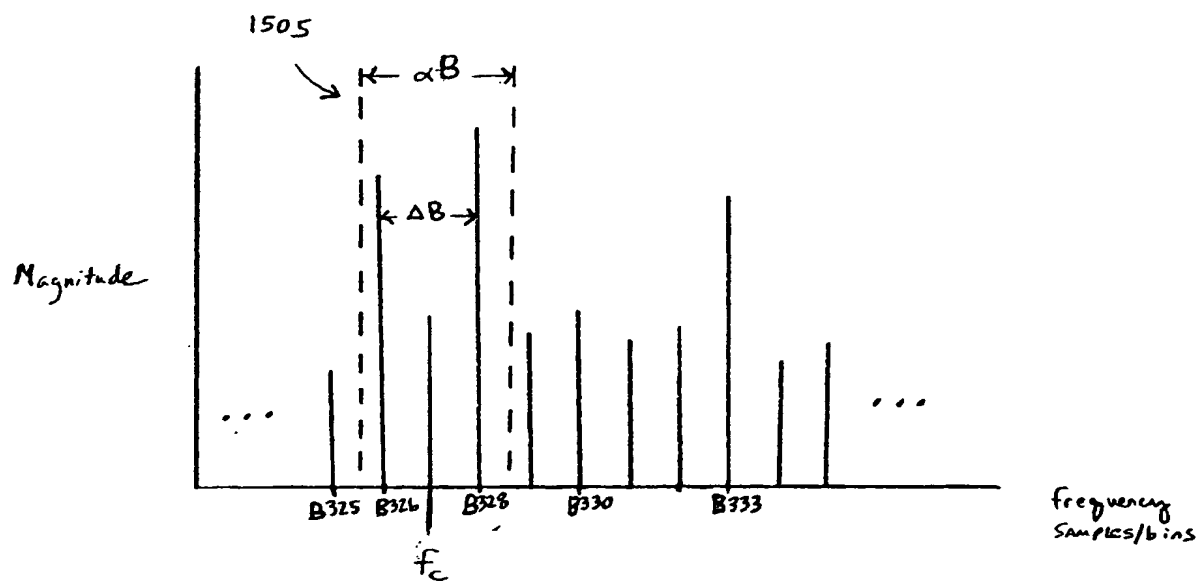


Figure 15

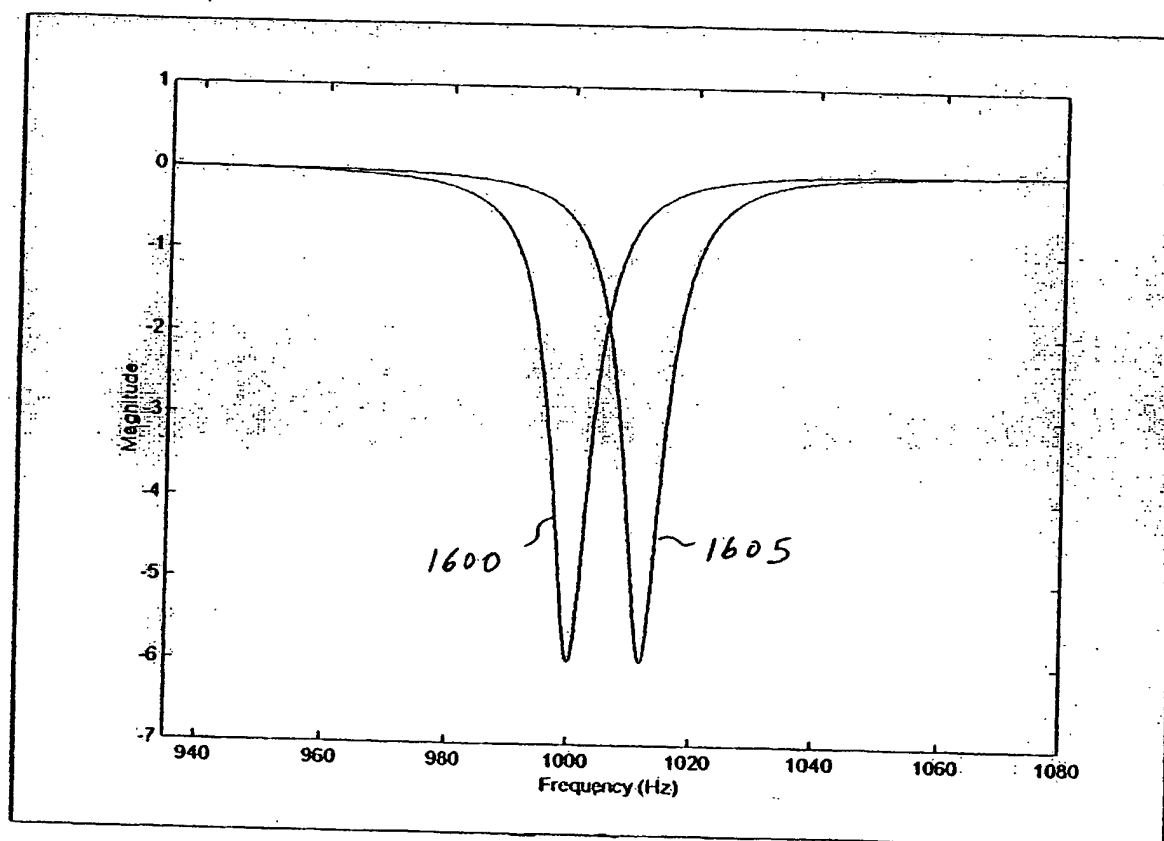


Figure 16



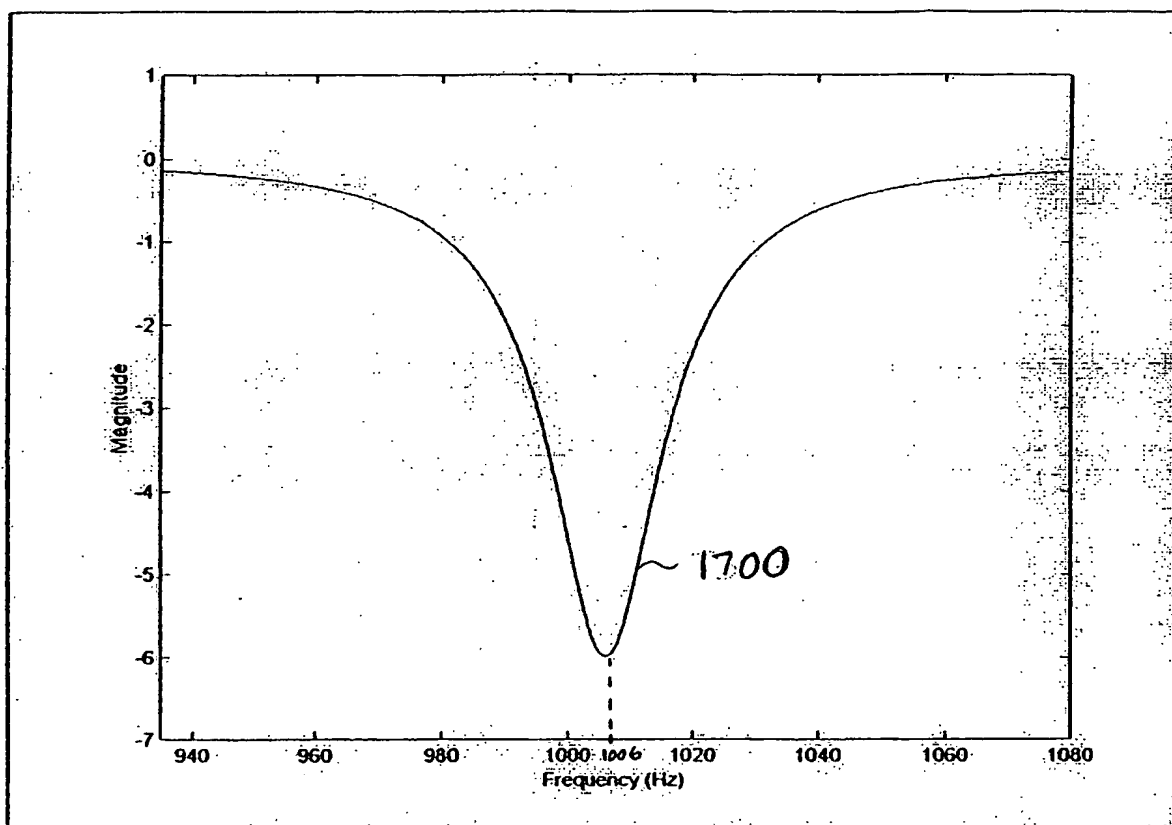


Figure 17

## INTERNATIONAL SEARCH REPORT

International Application No

PC1/US 03/07943

## A. CLASSIFICATION OF SUBJECT MATTER

IPC 7 H04R3/02 G10L19/02 //H04R5/04, H04B15/00

According to International Patent Classification (IPC) or to both national classification and IPC

## B. FIELDS SEARCHED

Minimum documentation searched (classification system followed by classification symbols)

IPC 7 H04R G10L

Documentation searched other than minimum documentation to the extent that such documents are included in the fields searched

Electronic data base consulted during the international search (name of data base and, where practical, search terms used)

EPO-Internal, WPI Data

## C. DOCUMENTS CONSIDERED TO BE RELEVANT

Category *	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.
X	US 6 058 194 A (GULLI CHRISTIAN ET AL) 2 May 2000 (2000-05-02)	1,2, 6-10,16, 17, 21-24, 33-42, 45-50
A	the whole document	3-5, 11-15, 18-20, 25-32, 43,44
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Date of the actual completion of the international search

4 July 2003

Date of mailing of the international search report

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# INTERNATIONAL SEARCH REPORT

International Application No  
PCT/US 03/07943

C.(Continuation) DOCUMENTS CONSIDERED TO BE RELEVANT		
Category *	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.
X	WO 01 97212 A (KENWOOD GEOBIT CORP ;KENWOOD CORP (JP); SATO YASUSHI (JP)) 20 December 2001 (2001-12-20)	1,2, 6-13, 15-17, 21-42, 45-50
A	page 2, line 15 -page 6, line 22 abstract	3-5,14, 18-20, 43,44
X	DE 198 14 180 C (SIEMENS AUDIOLOGISCHE TECHNIK) 7 October 1999 (1999-10-07)	1,8,16, 23
A	the whole document	2-7, 9-15, 17-22, 24-50

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